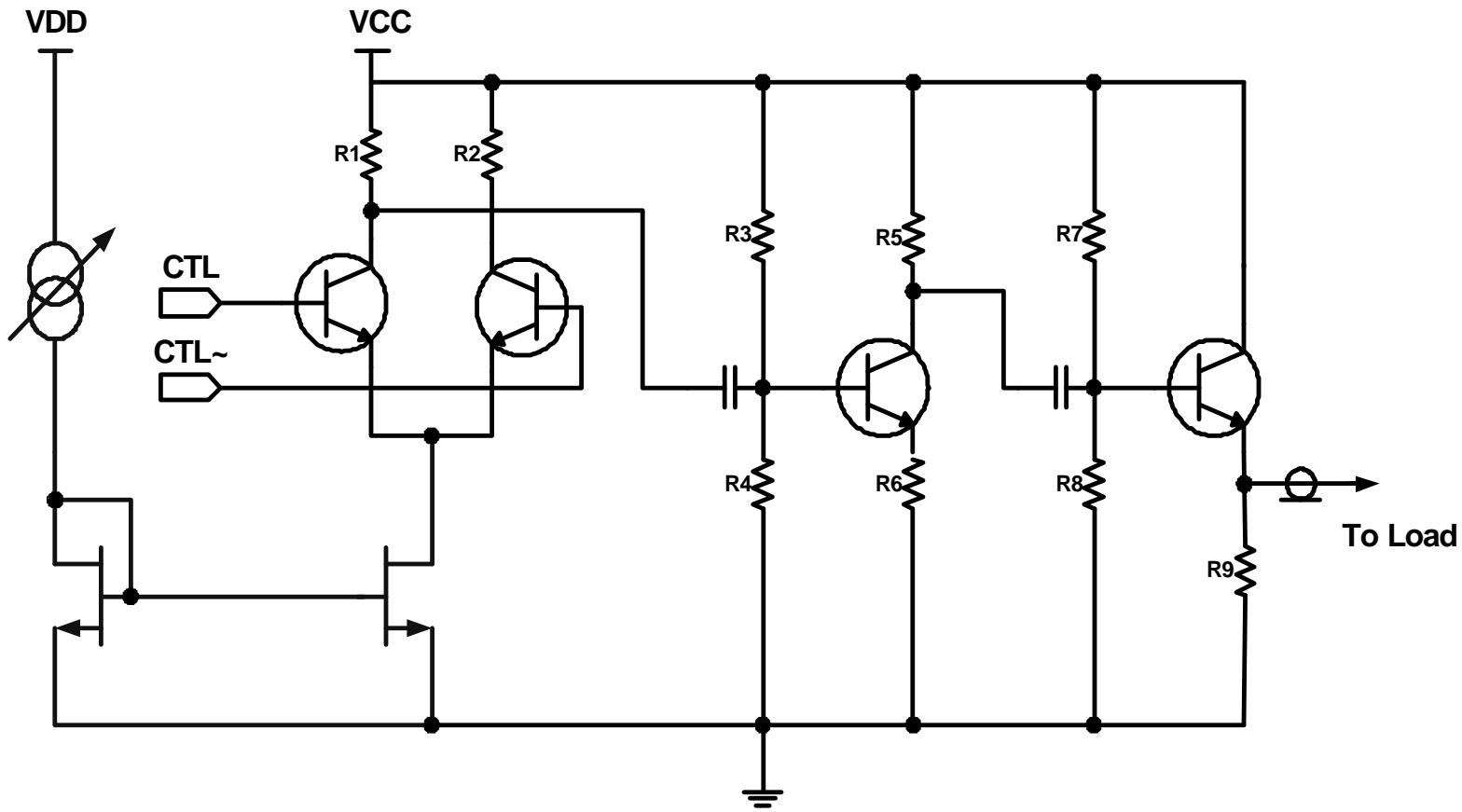
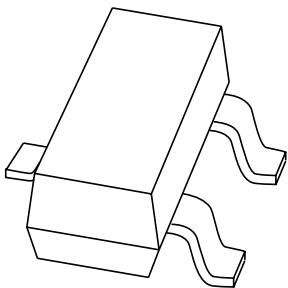




Proposed Circuit for Calibration



DATA SHEET



BFR540 NPN 9 GHz wideband transistor

Product specification
Supersedes data of 1999 Aug 23

2000 May 30

NPN 9 GHz wideband transistor**BFR540****FEATURES**

- High power gain
- Low noise figure
- High transition frequency
- Gold metallization ensures excellent reliability.

APPLICATIONS

RF front end wideband applications in the GHz range, such as analog and digital cellular telephones, cordless telephones (CT1, CT2, DECT, etc.), radar detectors, satellite TV tuners (SATV), MATV/CATV amplifiers and repeater amplifiers in fibre-optic systems.

DESCRIPTION

NPN silicon planar epitaxial transistor in a SOT23 plastic package.

PINNING

PIN	DESCRIPTION
1	base
2	emitter
3	collector

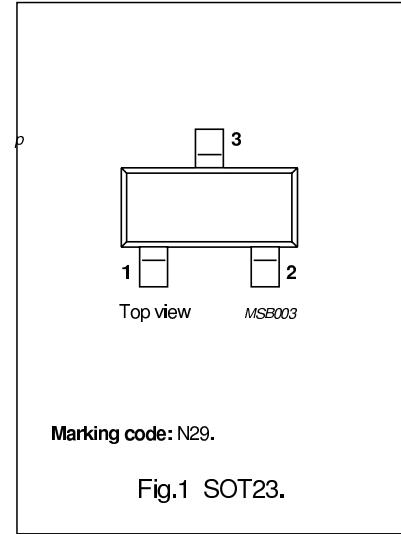


Fig.1 SOT23.

QUICK REFERENCE DATA

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
V_{CBO}	collector-base voltage	open emitter	—	—	20	V
V_{CES}	collector-emitter voltage	$R_{BE} = 0$	—	—	15	V
I_C	DC collector current		—	—	120	mA
P_{tot}	total power dissipation	$T_s \leq 70^\circ\text{C}$; note 1	—	—	500	mW
h_{FE}	DC current gain	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}$	100	120	250	
C_{re}	feedback capacitance	$I_C = i_c = 0; V_{CB} = 8 \text{ V}; f = 1 \text{ MHz}$	—	0.6	—	pF
f_T	transition frequency	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; f = 1 \text{ GHz}$	—	9	—	GHz
G_{UM}	maximum unilateral power gain	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	—	14	—	dB
		$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 2 \text{ GHz}$	—	7	—	dB
$ s_{21} ^2$	insertion power gain	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	12	13	—	dB
F	noise figure	$\Gamma_s = \Gamma_{opt}; I_C = 10 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	—	1.3	1.8	dB
		$\Gamma_s = \Gamma_{opt}; I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	—	1.9	2.4	dB
		$\Gamma_s = \Gamma_{opt}; I_C = 10 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 2 \text{ GHz}$	—	2.1	—	dB

Note

1. T_s is the temperature at the soldering point of the collector tab.

NPN 9 GHz wideband transistor

BFR540

LIMITING VALUES

In accordance with the Absolute Maximum System (IEC 60134).

SYMBOL	PARAMETER	CONDITIONS	MIN.	MAX.	UNIT
V_{CBO}	collector-base voltage	open emitter	–	20	V
V_{CES}	collector-emitter voltage	$R_{BE} = 0$	–	15	V
V_{EBO}	emitter-base voltage	open collector	–	2.5	V
I_C	DC collector current		–	120	mA
P_{tot}	total power dissipation	$T_s \leq 70^\circ\text{C}$; note 1	–	500	mW
T_{stg}	storage temperature		–65	150	°C
T_j	junction temperature		–	175	°C

Note

1. T_s is the temperature at the soldering point of the collector tab.

THERMAL CHARACTERISTICS

SYMBOL	PARAMETER	CONDITIONS	VALUE	UNIT
$R_{th\ j-s}$	from junction to soldering point	see note 1	260	K/W

Note

1. T_s is the temperature at the soldering point of the collector tab.

NPN 9 GHz wideband transistor

BFR540

CHARACTERISTICS

 $T_j = 25^\circ\text{C}$ unless otherwise specified.

SYMBOL	PARAMETER	CONDITIONS	MIN.	TYP.	MAX.	UNIT
I_{CBO}	collector cut-off current	$I_E = 0; V_{CB} = 8 \text{ V}$	—	—	50	nA
h_{FE}	DC current gain	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}$	100	120	250	
C_e	emitter capacitance	$I_C = i_e = 0; V_{EB} = 0.5 \text{ V}; f = 1 \text{ MHz}$	—	2	—	pF
C_c	collector capacitance	$I_E = i_e = 0; V_{CB} = 8 \text{ V}; f = 1 \text{ MHz}$	—	0.9	—	pF
C_{re}	feedback capacitance	$I_C = 0; V_{CB} = 8 \text{ V}; f = 1 \text{ MHz}$	—	0.6	—	pF
f_T	transition frequency	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; f = 1 \text{ GHz}$	—	9	—	GHz
G_{UM}	maximum unilateral power gain; note 1	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	—	14	—	dB
		$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 2 \text{ GHz}$	—	7	—	dB
$ s_{21} ^2$	insertion power gain	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	12	13	—	dB
F	noise figure	$\Gamma_s = \Gamma_{opt}; I_C = 10 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	—	1.3	1.8	dB
		$\Gamma_s = \Gamma_{opt}; I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	—	1.9	2.4	dB
		$\Gamma_s = \Gamma_{opt}; I_C = 10 \text{ mA}; V_{CE} = 8 \text{ V}; T_{amb} = 25^\circ\text{C}; f = 2 \text{ GHz}$	—	2.1	—	dB
P_{L1}	output power at 1 dB gain compression	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; R_L = 50 \Omega; T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz}$	—	21	—	dBm
ITO	third order intercept point	note 2	—	34	—	dBm
V_o	output voltage; note 3	$I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; Z_L = Z_S = 75 \Omega; T_{amb} = 25^\circ\text{C}$	—	550	—	mV

Notes

1. G_{UM} is the maximum unilateral power gain, assuming s_{12} is zero and

$$G_{UM} = 10 \log \frac{|s_{21}|^2}{(1 - |s_{11}|^2)(1 - |s_{22}|^2)} \text{ dB.}$$

2. $I_C = 40 \text{ mA}; V_{CE} = 8 \text{ V}; R_L = 50 \Omega;$
 $T_{amb} = 25^\circ\text{C}; f = 900 \text{ MHz};$
 $f_p = 900 \text{ MHz}; f_q = 902 \text{ MHz};$
measured at $f_{(2p-q)} = 898 \text{ MHz}$ and $f_{(2q-p)} = 904 \text{ MHz}$.
3. $d_{im} = -60 \text{ dB}$ (DIN 45004B);
 $V_p = V_o; V_q = V_o - 6 \text{ dB}; f_p = 795.25 \text{ MHz};$
 $V_R = V_o - 6 \text{ dB}; f_q = 803.25 \text{ MHz}; f_r = 805.25 \text{ MHz};$
measured at $f_{(p+q+r)} = 793.25 \text{ MHz}$; preliminary data.

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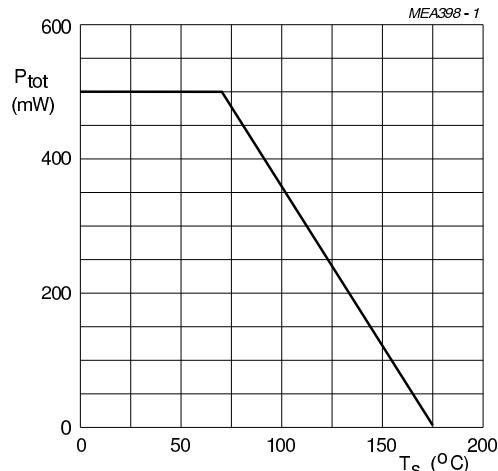


Fig.2 Power derating curve.

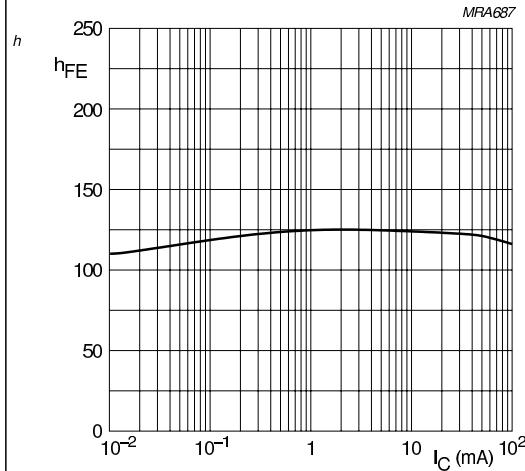
 $V_{CE} = 8$ V.

Fig.3 DC current gain as a function of collector current.

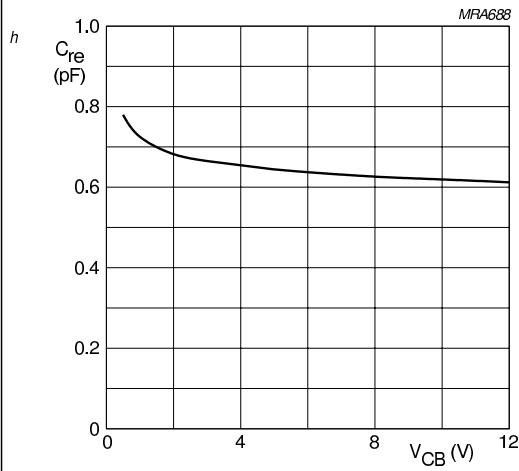
 $I_C = 0$; $f = 1$ MHz.

Fig.4 Feedback capacitance as a function of collector-base voltage.

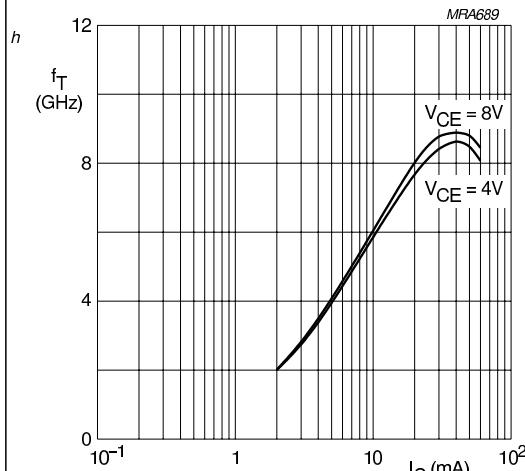
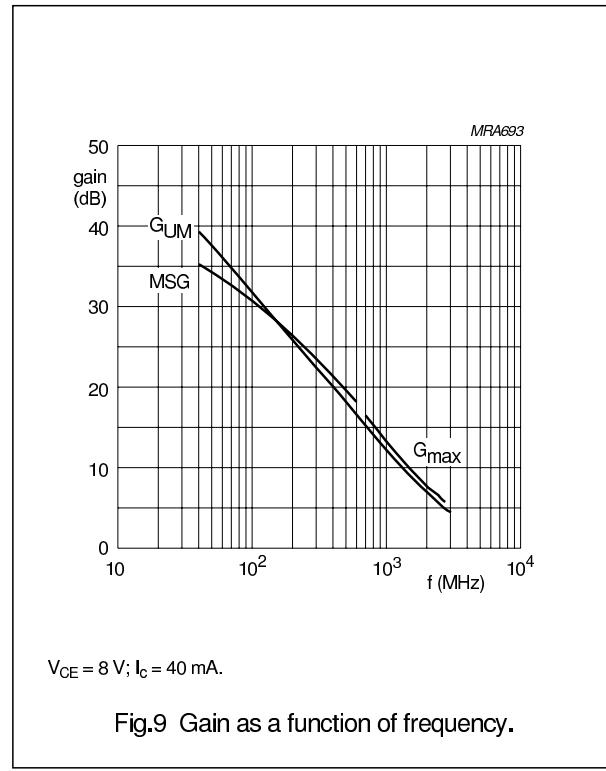
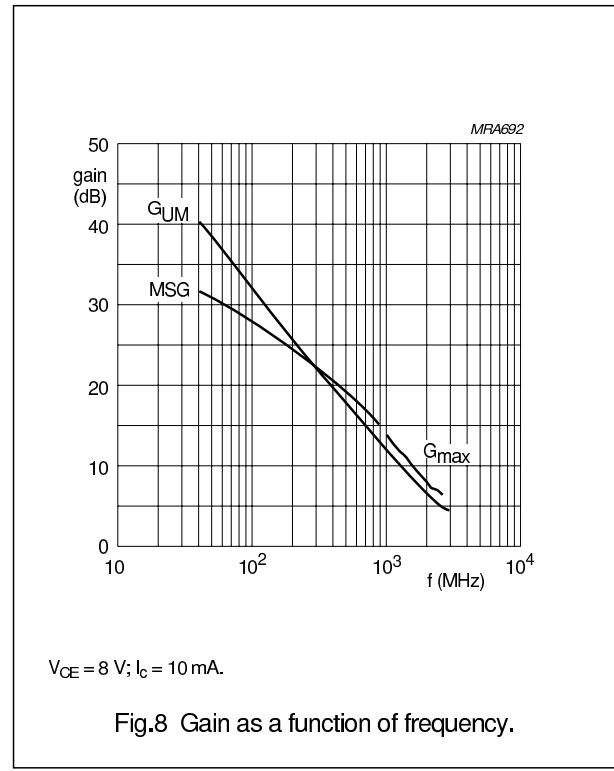
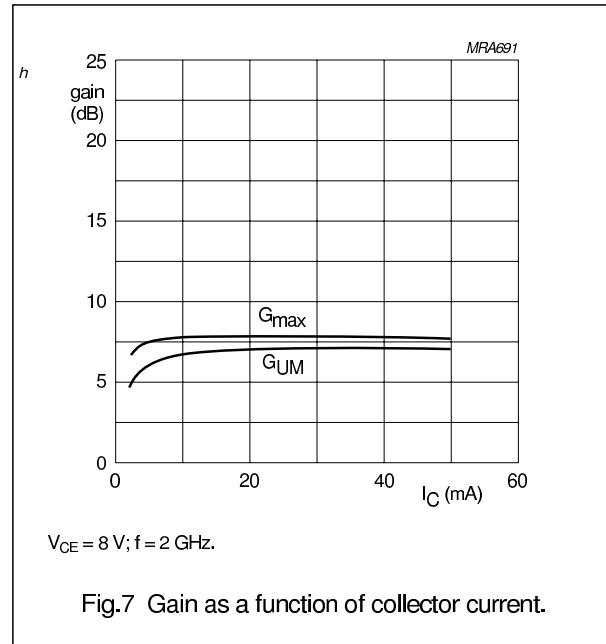
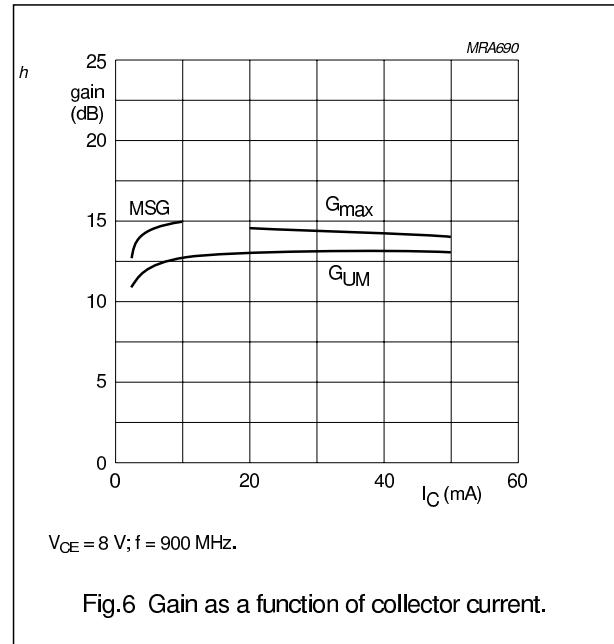
 $T_{amb} = 25$ $^{\circ}$ C; $f = 1$ GHz.

Fig.5 Transition frequency as a function of collector current.

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In Figs 6 to 9, G_{UM} = maximum unilateral power gain;
 MSG = maximum stable gain; G_{max} = maximum available gain.



NPN 9 GHz wideband transistor

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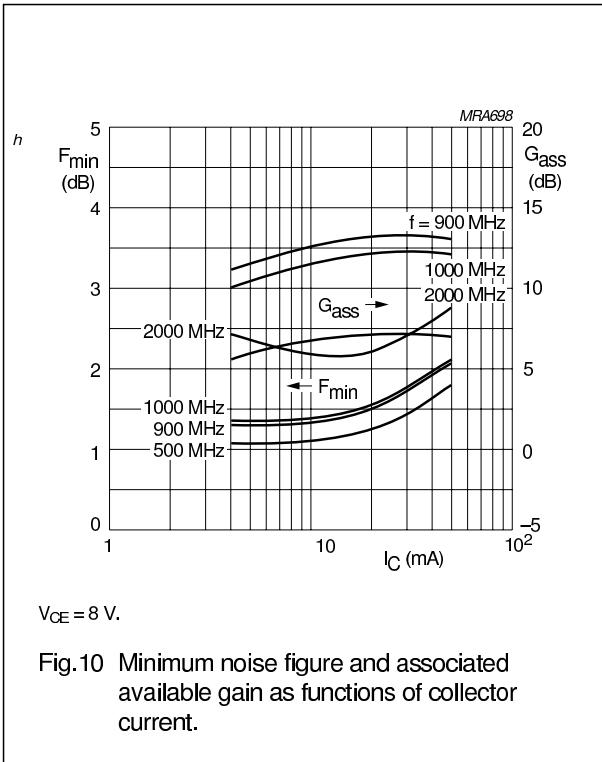
 $V_{CE} = 8$ V.

Fig.10 Minimum noise figure and associated available gain as functions of collector current.

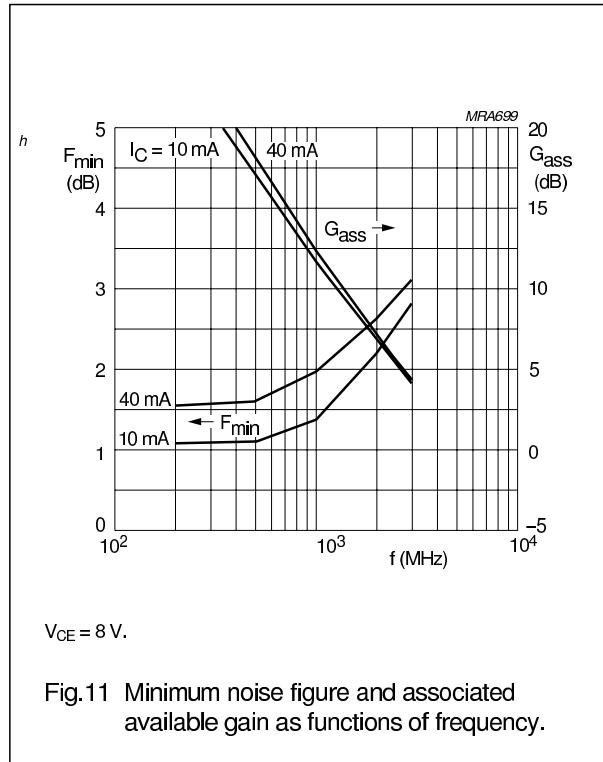
 $V_{CE} = 8$ V.

Fig.11 Minimum noise figure and associated available gain as functions of frequency.

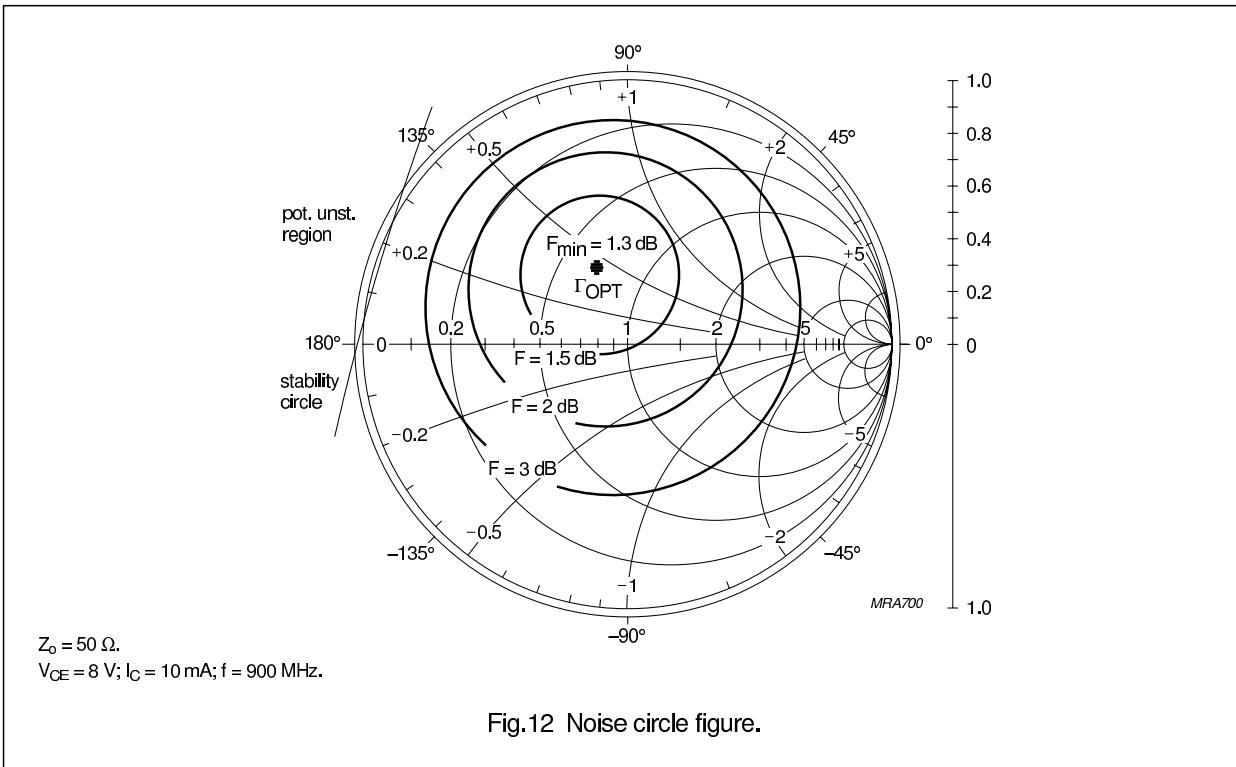
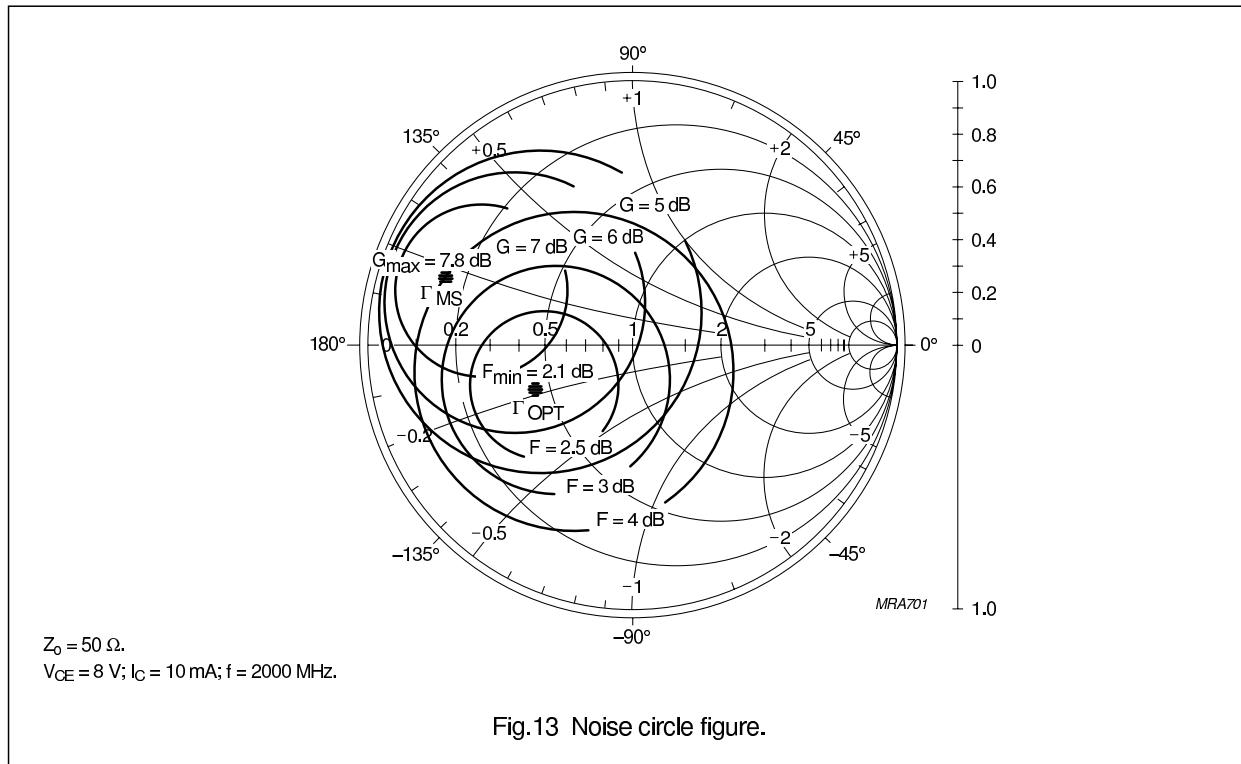


Fig.12 Noise circle figure.

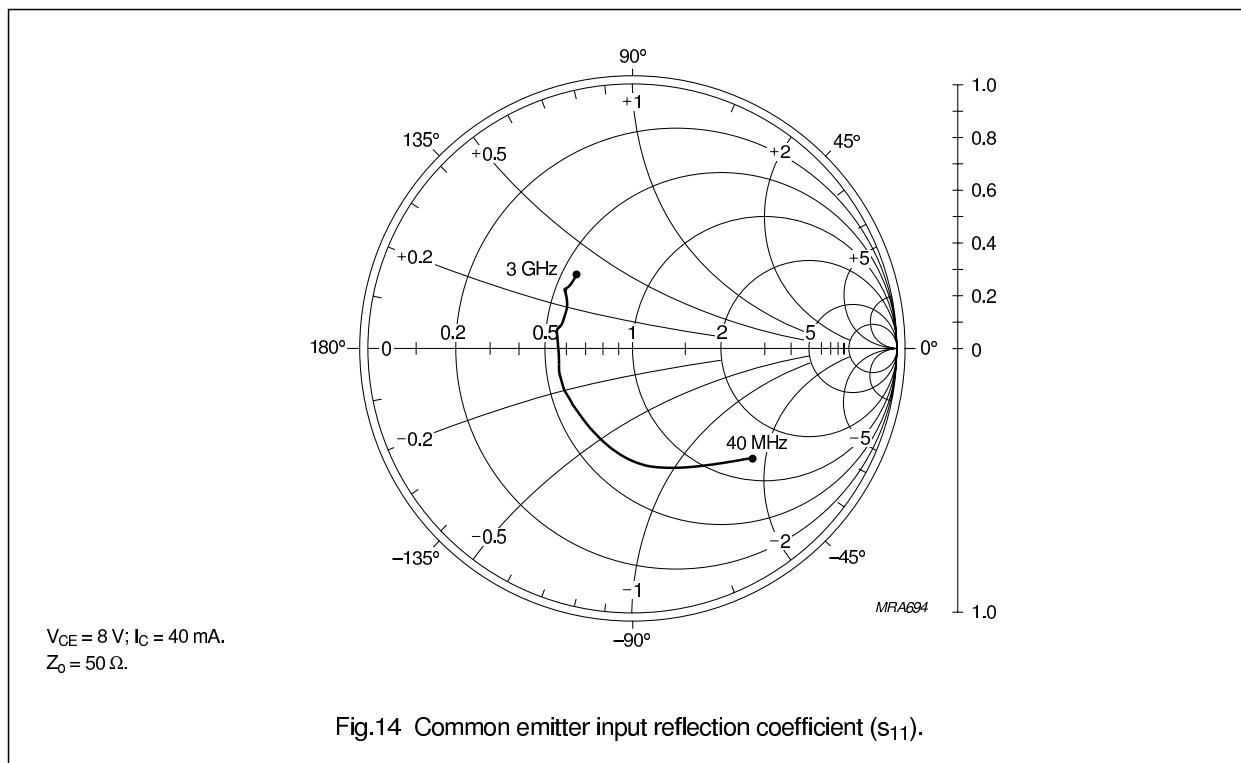
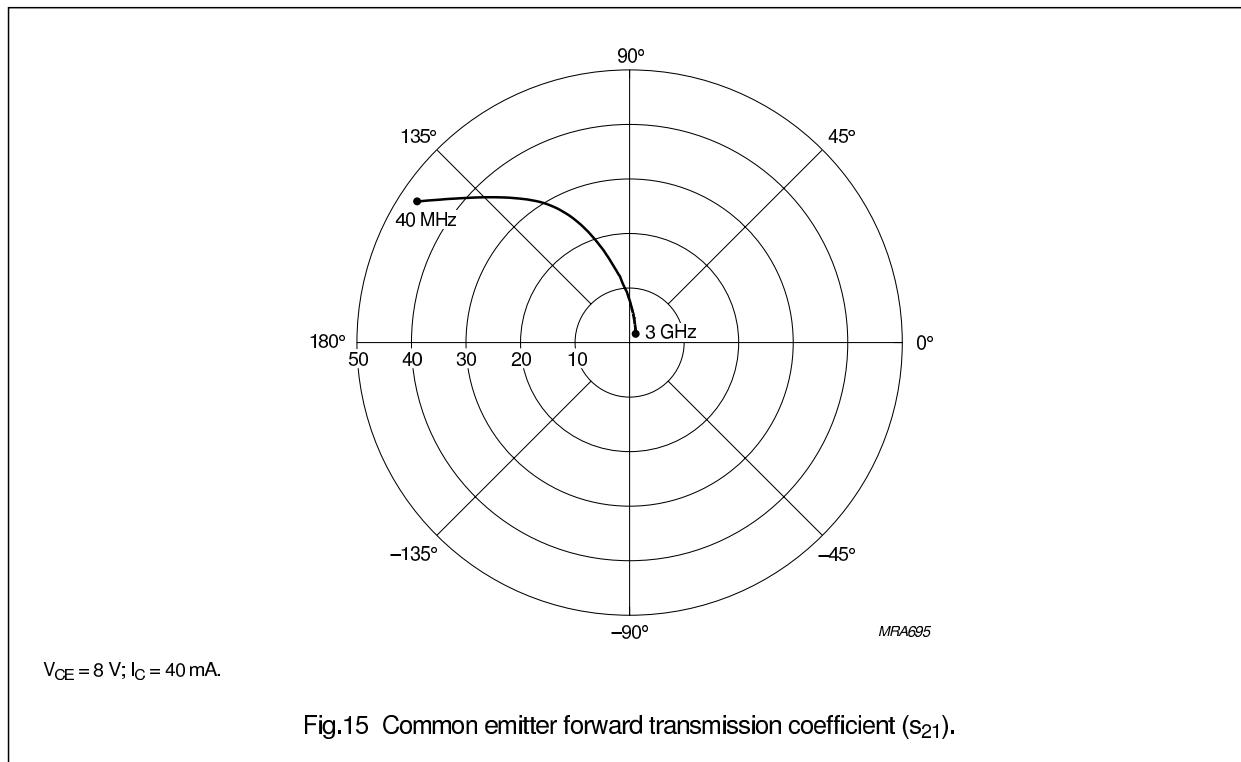
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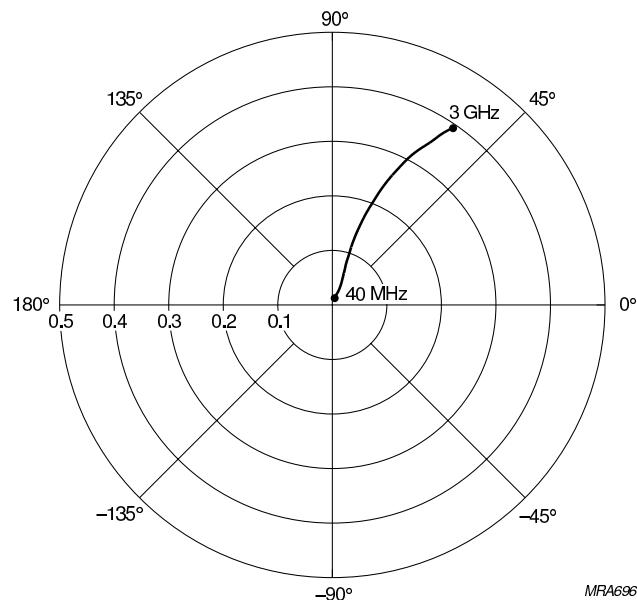
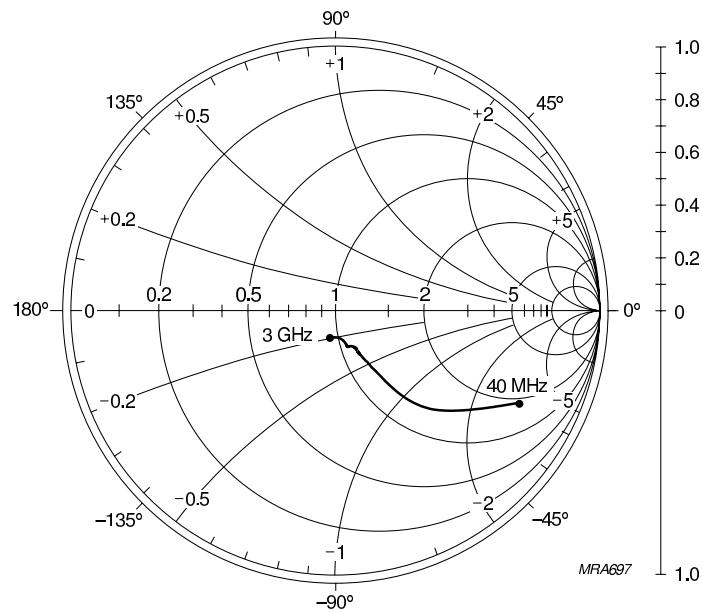
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Fig.14 Common emitter input reflection coefficient (s_{11}).Fig.15 Common emitter forward transmission coefficient (s_{21}).

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 $V_{CE} = 8 \text{ V}; I_C = 40 \text{ mA}.$ Fig.16 Common emitter reverse transmission coefficient (s_{12}). $V_{CE} = 8 \text{ V}; I_C = 40 \text{ mA}.$
 $Z_0 = 50 \Omega.$ Fig.17 Common emitter output reflection coefficient (s_{22}).

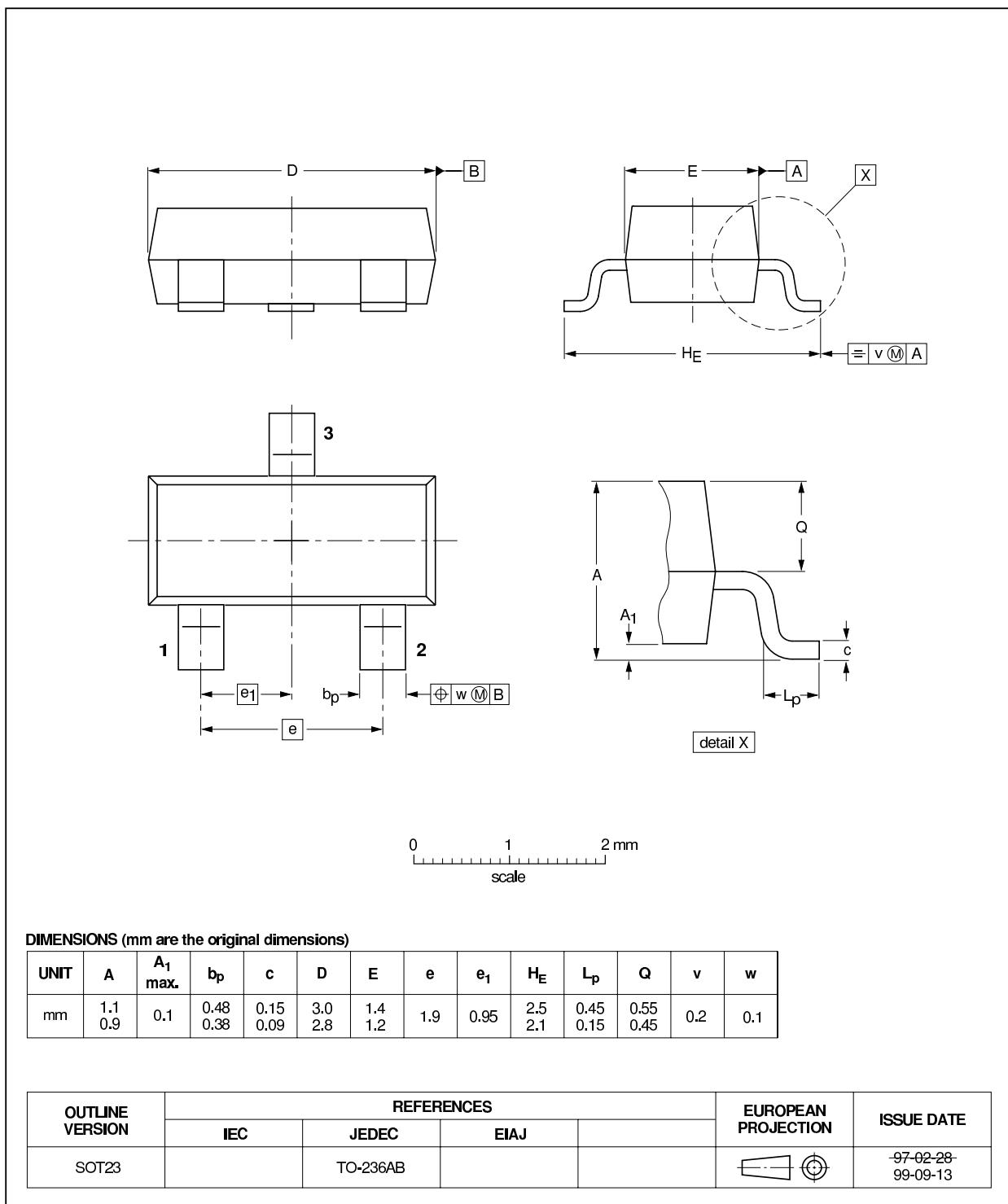
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BFR540

PACKAGE OUTLINE

Plastic surface mounted package; 3 leads

SOT23



NPN 9 GHz wideband transistor**BFR540****DATA SHEET STATUS**

DATA SHEET STATUS	PRODUCT STATUS	DEFINITIONS ⁽¹⁾
Objective specification	Development	This data sheet contains the design target or goal specifications for product development. Specification may change in any manner without notice.
Preliminary specification	Qualification	This data sheet contains preliminary data, and supplementary data will be published at a later date. Philips Semiconductors reserves the right to make changes at any time without notice in order to improve design and supply the best possible product.
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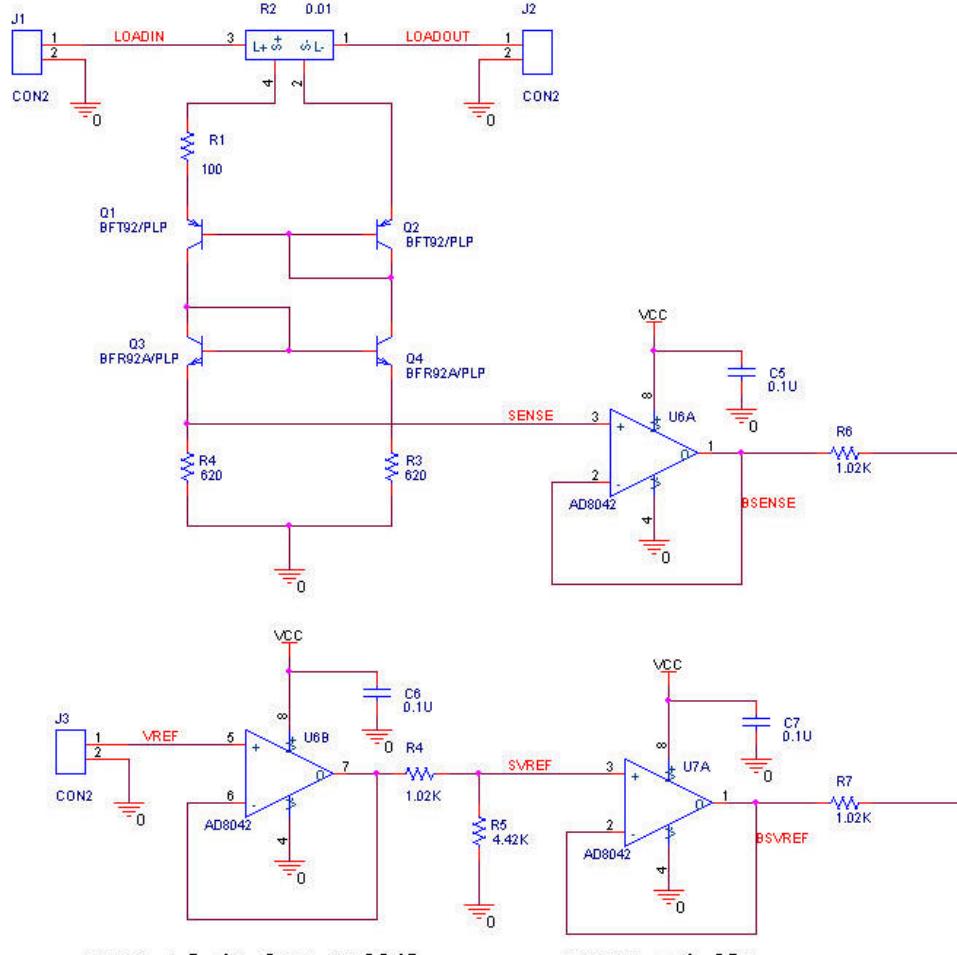


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Circuit for CSC Current Sensing

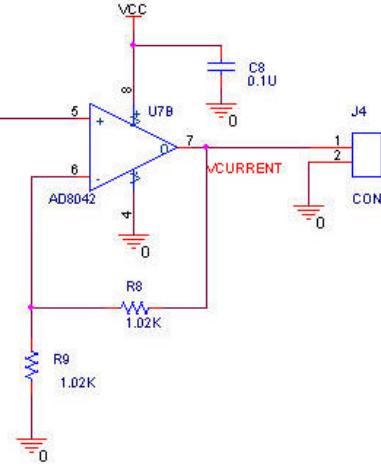
$$V_{sense} = I_{load} \cdot (R_2/R_1) \cdot R_4$$



V_{REF} : 2.4V from AD9042

$SVREF$ = 1.95V

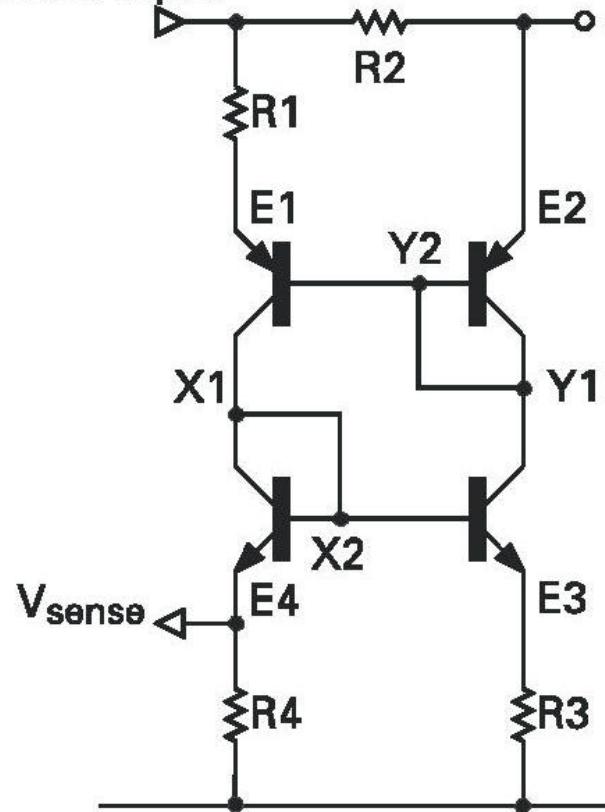
$$I_{load} = (V_{current} - SVREF) / (R_2/R_1) \cdot R_4$$





Principle of Operation

Current Input



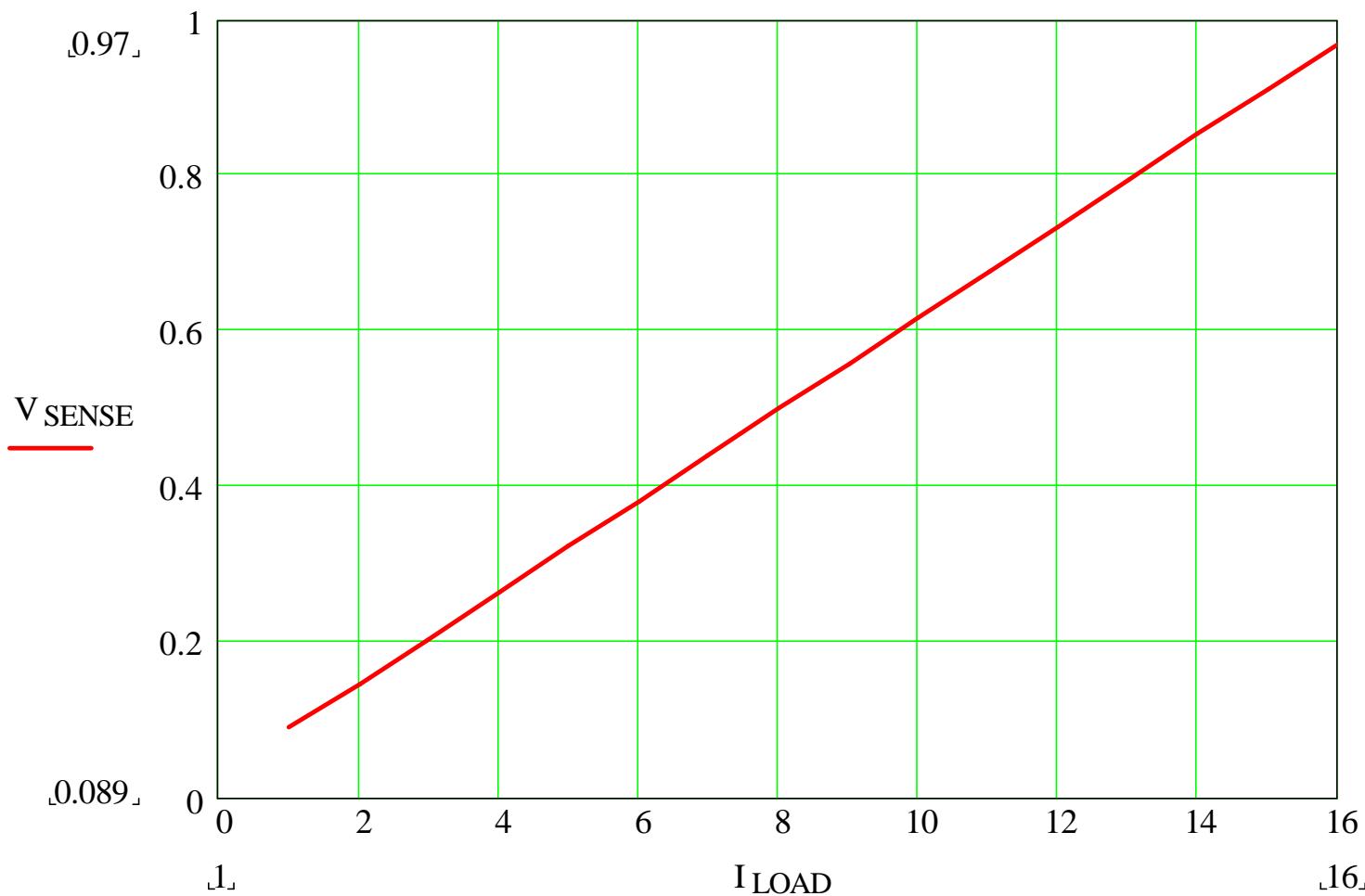
$$V_{BE_Q1} = V_{BE_Q2}$$

Where $V_{BE_Q2} = I_{LOAD} \bullet R_2$

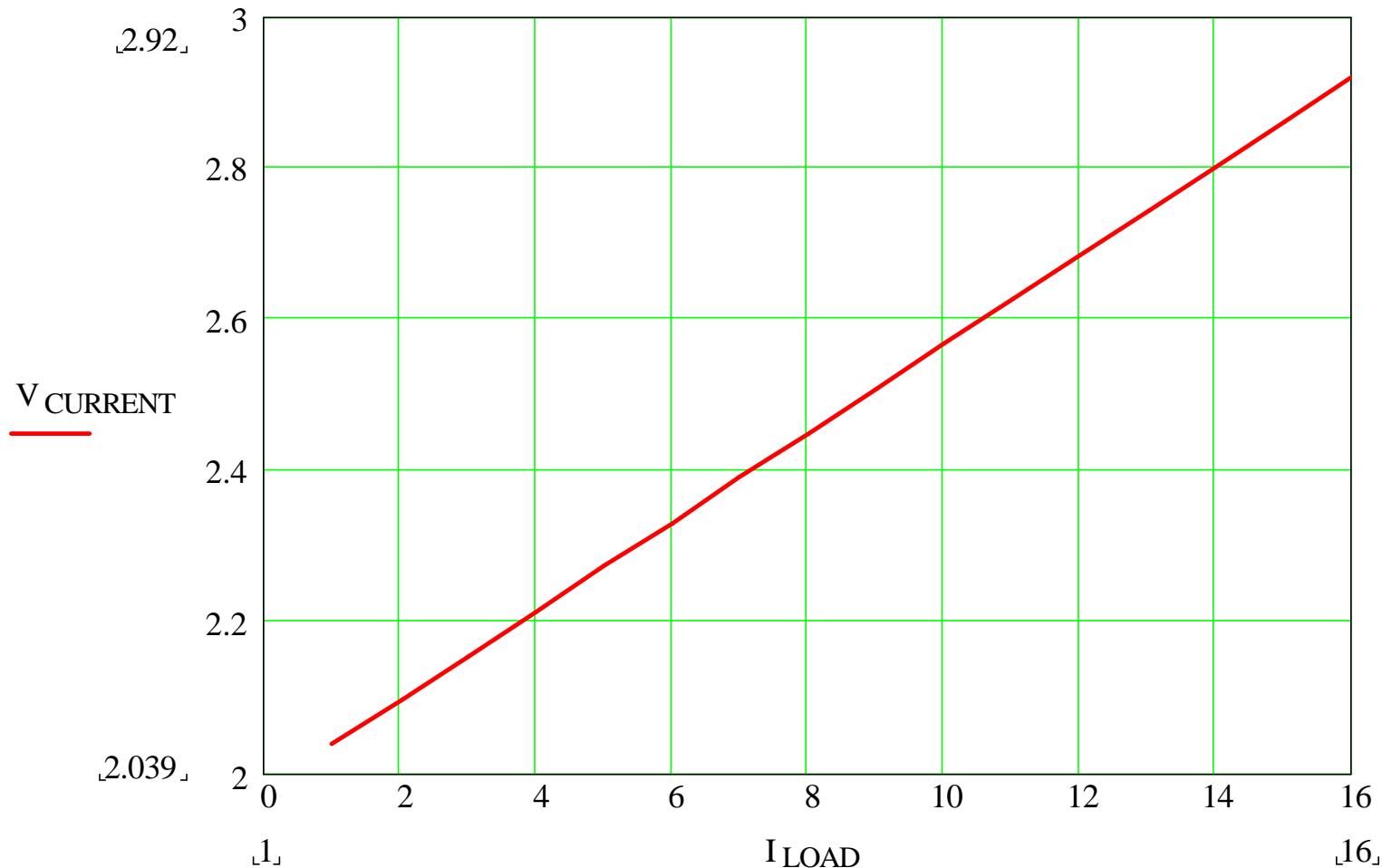
$$I_{R1} = \frac{V_{BE_Q2}}{R_1} = I_{LOAD} \bullet \frac{R_2}{R_1}$$

$$V_{SENSE} = I_{R1} \bullet R_4 = I_{LOAD} \bullet \frac{R_2}{R_1} \bullet R_4$$

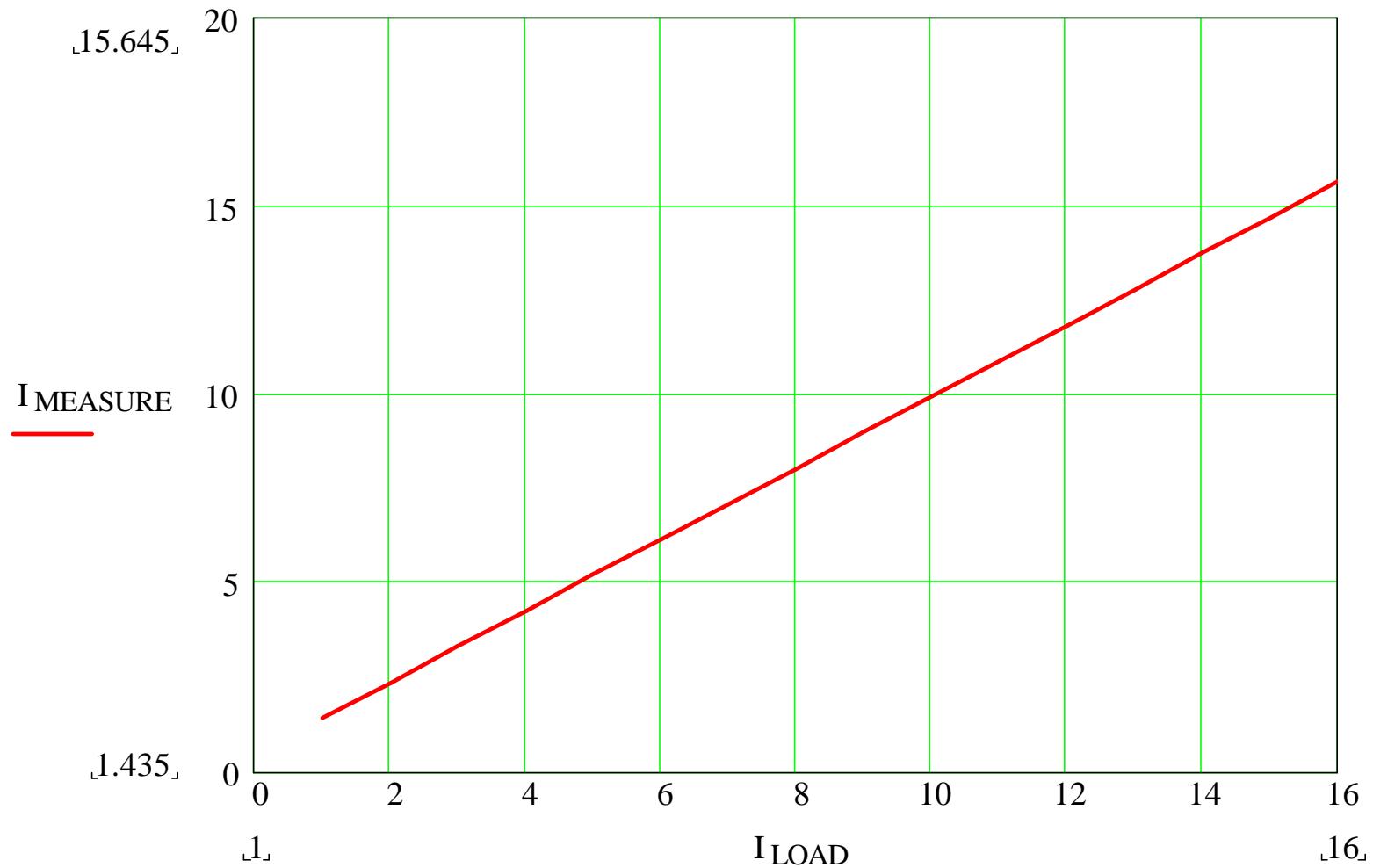
Vsense



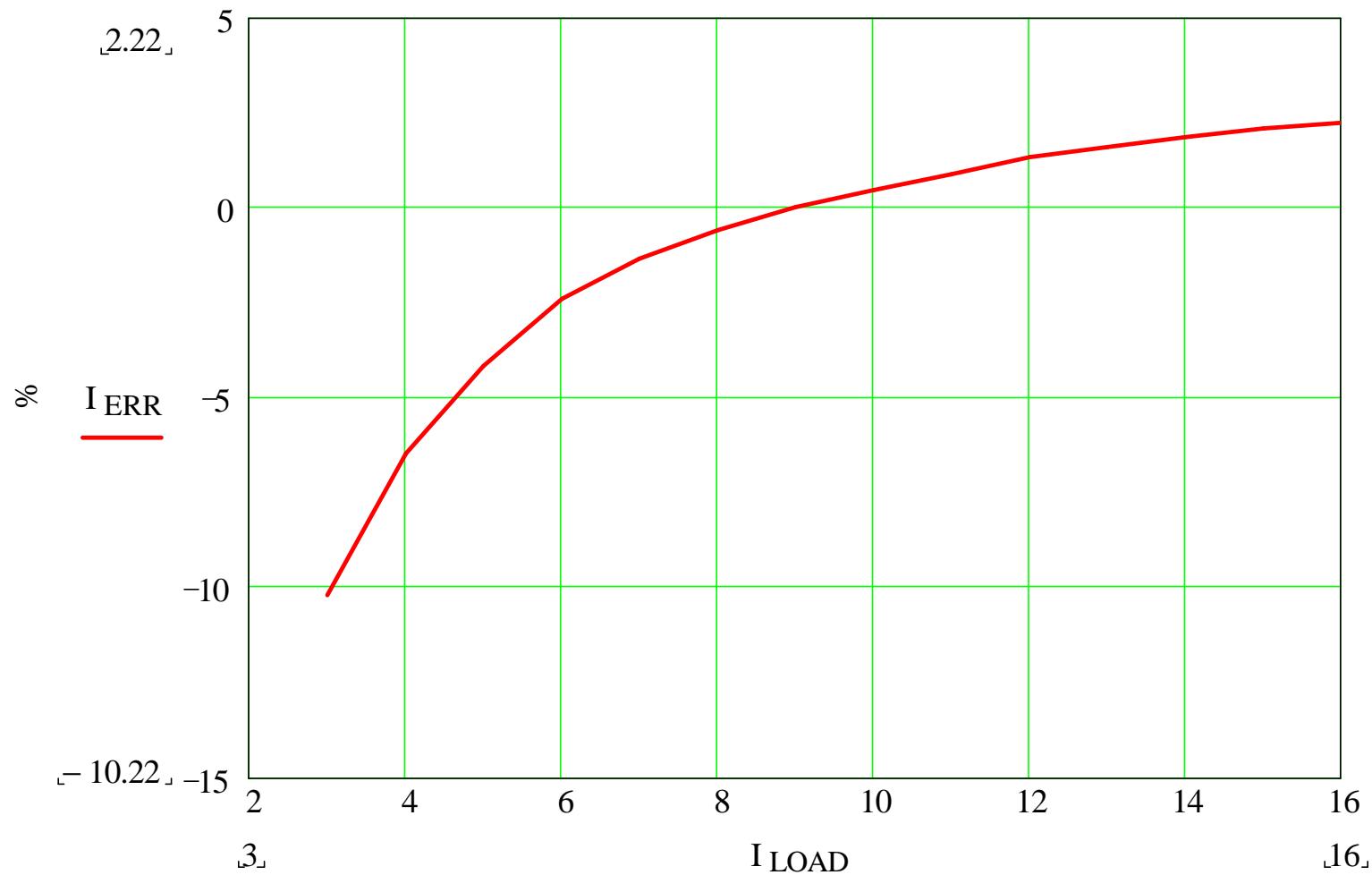
Vcurrent



Imeasure

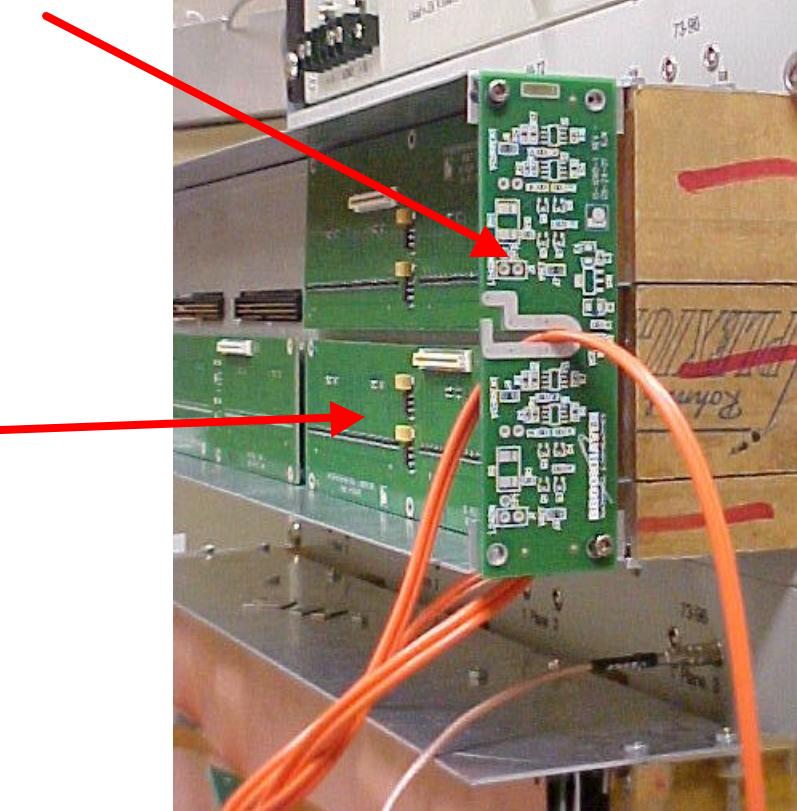


Error



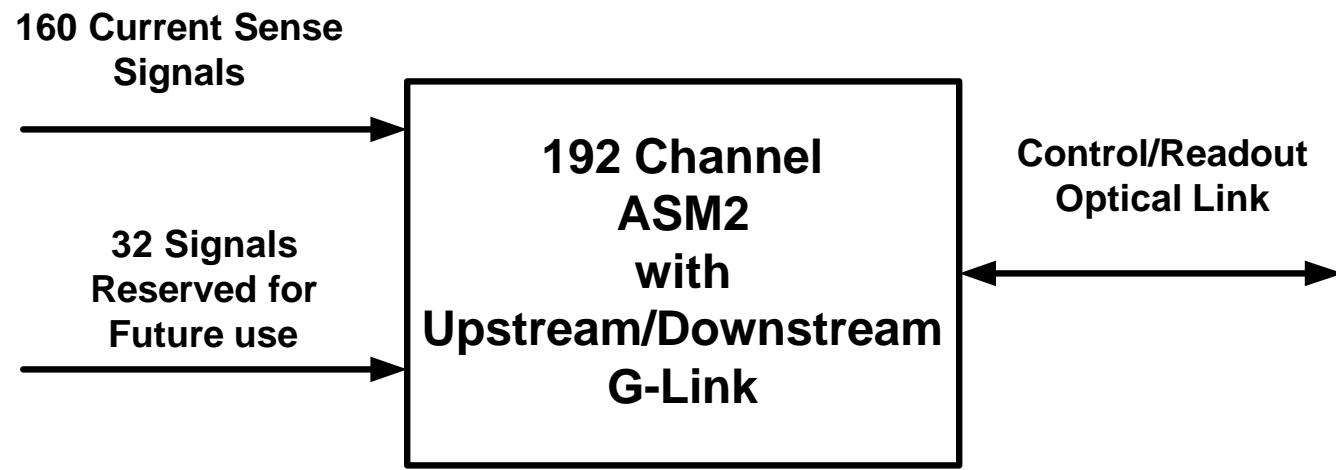
Location of Current Sensing Board

Current Sense Board



ASM1 Boards

Current Sense Readout



SM-8 COMPLEMENTARY CURRENT MIRROR

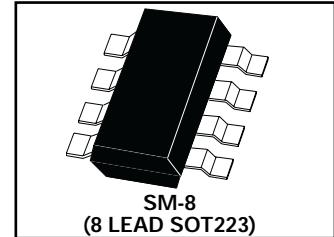
DESCRIPTION

The ZDS1009 current mirror has been developed specifically for high side, current sense plus level translation applications and as such will find a broad applications base including battery charge management, DC motor control and over current monitoring functions. It is of particular interest for current sense applications for feedback purposes in fast battery chargers for Li-Ion cell based systems.

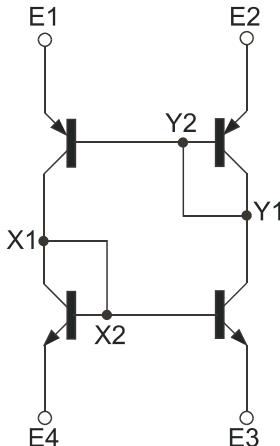
The device functions by sensing the voltage developed across an external (user defined) high side current sense resistor, and by an arrangement of current mirrors refer this sensed voltage, with or without multiplication, to a low side referenced signal. This signal can then be used, for example, to close the control loop to a controller IC, for a DC-DC converter providing charge to a battery.

FEATURES

- Excellent Temperature Tracking Characteristics
- Compact Cost Effective Solution
- Simplifies Circuit Implementation
- Broad application base from Single Cell Li-ion High Side Current sense chargers to Multi-cell Lead-Acid systems
- Only 4 Connections required



SCHEMATIC DIAGRAM

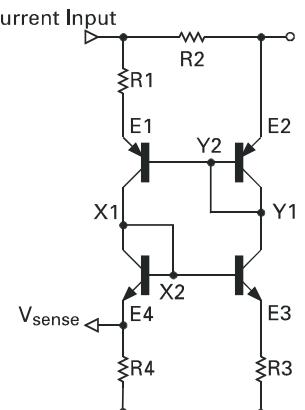


TYPICAL APPLICATION CIRCUIT

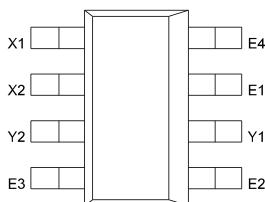
$$V_{\text{sense}} = IR_2 \frac{R_4}{R_1}$$

For balance $R_3=R_4$
eg
 $R_2=100\text{m}\Omega$
 $R_1=R_3=R_4=100\Omega$

V_{sense} sensitivity = 100mV/A



CONNECTION DIAGRAM



ZDS1009

ABSOLUTE MAXIMUM RATINGS.

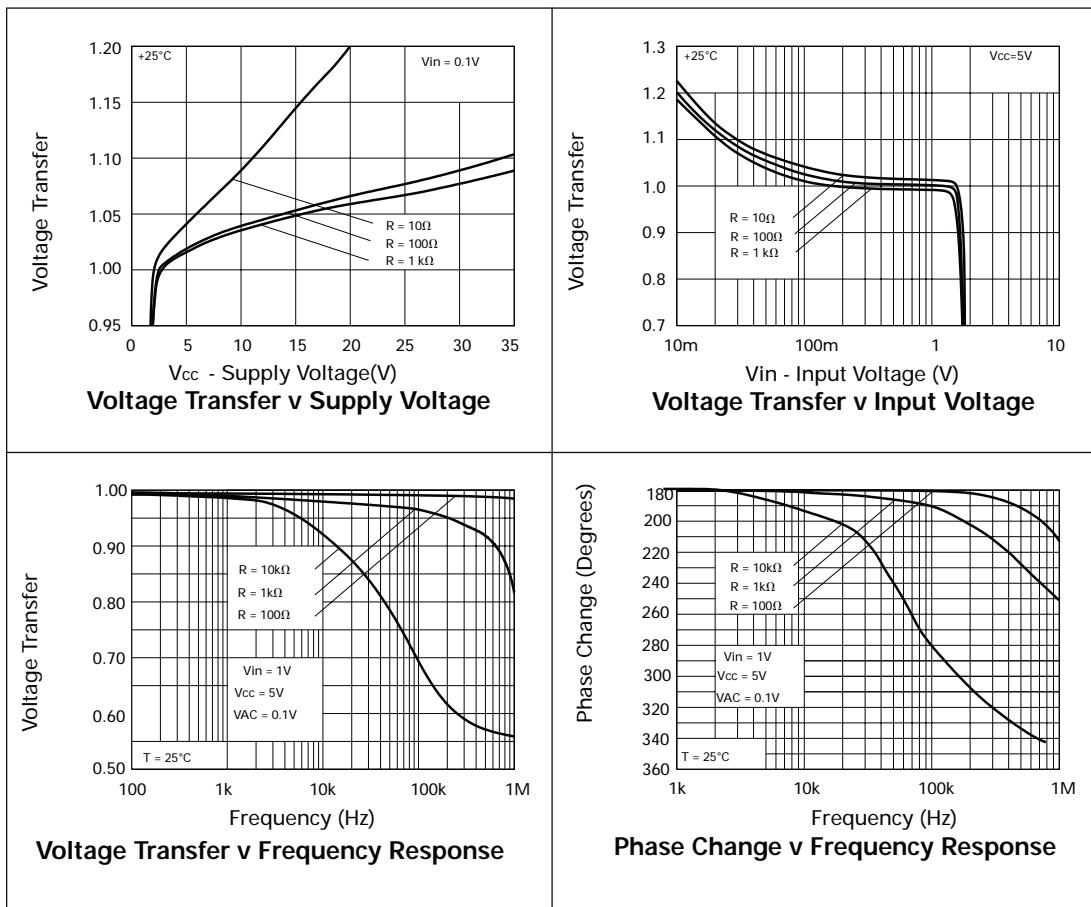
PARAMETER	SYMBOL	VALUE	UNIT
Maximum Operating Voltage	V_{y1-x1}	120	V
Maximum Voltage (E1-E2,E3-E4)	$V_{E-E'}$	10	V
Peak Pulse Current	I_M	4	A
Continuous Current (E1-E4,E2-E3)	I_C	1	A
Total Power Dissipation at $T_{amb} = 25^\circ C^*$	P_{tot}	2	W
Operating and Storage Temperature Range	$T_J; T_{stg}$	-55 to +150	°C

* The power which can be dissipated assuming the device is mounted in a typical manner on a PCB with copper equal to 2 inches square.

ELECTRICAL CHARACTERISTICS (at $T_{amb}=25^\circ C$)

Parameter	Symbol	Min	Max	Unit	Conditions
Breakdown Voltage	BV_{Y1-X1}	120		V	$I_{Y1}=100\mu A$
Breakdown Voltage	BV_{X1-E1}	-30		V	$I_{X1}=-10mA$
Breakdown Voltage	BV_{Y1-E3}	30		V	$I_{Y1}=10mA$
Breakdown Voltage	BV_{E1-Y1}	-12		V	$I_{E1}=-100\mu A$
Breakdown Voltage	BV_{E2-Y1}	-6		V	$I_{E2}=-100\mu A$
Breakdown Voltage	BV_{E3-X1}	12		V	$I_{E3}=100\mu A$
Breakdown Voltage	BV_{E4-X1}	6		V	$I_{E4}=100\mu A$
Leakage	I_{Y1}		50	nA	$V_{Y1-X1}=100V$
Leakage	I_{X1}		-10	μA	$V_{X1-E1}=-30V, V_{y1}=V_{E1}$
Leakage	I_{Y1}		10	μA	$V_{Y1-E3}=30V, V_{X1}=V_{E3}$
Leakage	I_{E1}		-100	nA	$V_{E1-Y1}=-8V$
Leakage	I_{E2}		-100	nA	$V_{E2-Y1}=-4V$
Leakage	I_{E3}		100	nA	$V_{E3-X1}=8V$
Leakage	I_{E4}		100	nA	$V_{E4-X1}=4V$
Input Voltage	V_{Y1-E2}	-1.45	-1.65	V	$I_{Y1}=-1A$
Input Voltage	V_{Y1-E3}	1.45	1.75	V	$I_{Y1}=1A, V_{X1}=V_{Y1}$
Input Voltage	V_{X1-E1}	-1.45	-1.75	V	$I_{X1}=-1A, V_{X1}=V_{Y1}$
Input Voltage	V_{X1-E4}	1.45	1.65	V	$I_{X1}=1A$
Transfer Characteristic	V_{OUT}	0.99	1.01	V	See Fig 1. $V_{CC}=5V$ $R1=R3=R4=100\Omega, V_{IN}=1V$
Transfer Characteristic	V_{OUT}	1		mV	See Fig 1. $V_{CC}=5V$ $R1=R3=R4=100\Omega, V_{IN}=5mV$
Output Zero-Offset Voltage	V_{OFFSET}		4	mV	See Fig 2. $V_{CC}=5V, R_2<1\Omega$ $R1=R3=R4=100\Omega$

TYPICAL CHARACTERISTICS



TEST CIRCUITS

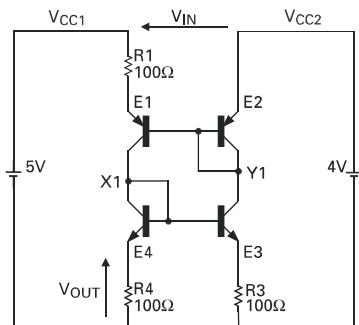


Figure 1
Transfer Characteristic Test Circuit

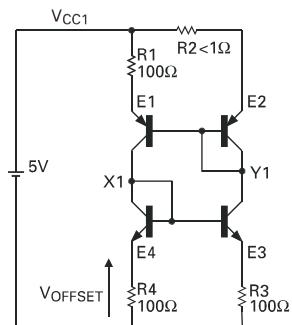
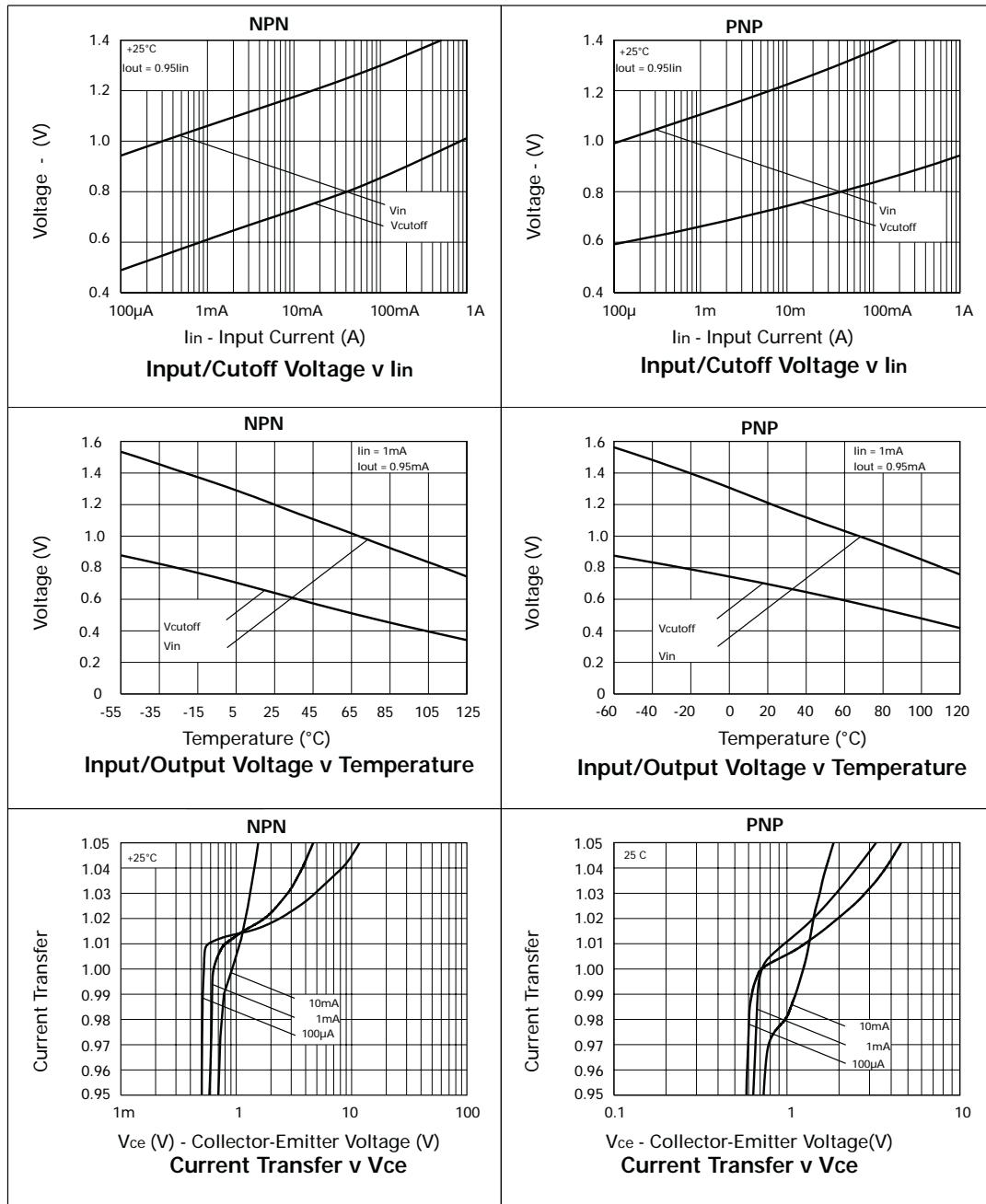


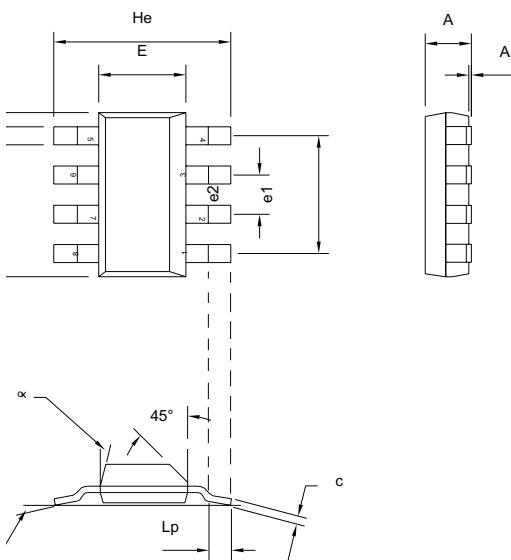
Figure 2
Output Zero-Offset Voltage Test Circuit

ZDS1009

TYPICAL CHARACTERISTICS



PACKAGE DIMENSIONS



DIM	Millimetres			Inches		
	Min	Typ	Max	Min	Typ	Max
A	-	-	1.7	-	-	0.067
A1	0.02	-	0.1	0.0008	-	0.004
b	-	0.7	-	-	0.028	-
c	0.24	-	0.32	0.009	-	0.013
D	6.3	-	6.7	0.248	-	0.264
E	3.3	-	3.7	0.130	-	0.145
e1	-	4.59	-	-	0.180	-
e2	-	1.53	-	-	0.060	-
He	6.7	-	7.3	0.264	-	0.287
Lp	0.9	-	-	0.035	-	-
α	-	-	15°	-	-	15°
β	-	10°	-	-	10°	-

ORDERING INFORMATION

DEVICE	PARTMARKING
ZDS1009	S1009



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